MP3389

12–String, Step-up White LED Driver

The Future of Analog IC Technology

DESCRIPTION

The MP3389 is a step-up controller with 12-channel current sources designed for driving the WLED arrays for large size LCD panel backlighting applications.

The MP3389 uses current mode, fixed frequency architecture. The switching frequency is programmable by an external frequency setting resistor. It drives an external MOSFET to boost up the output voltage from a 5V to 28V input supply. The MP3389 regulates the current in each LED string to the programmed value set by an external current setting resistor.

The MP3389 applies 12 internal current sources for current balance. And the current matching can achieve 2.5% regulation accuracy between strings. Its low 550mV regulation voltage on LED current sources reduces power loss and improves efficiency.

PWM dimming is implemented with external PWM input signal or DC input signal. The dimming PWM signal can be generated internally, and the dimming frequency is programmed by an external setting capacitor.

FEATURES

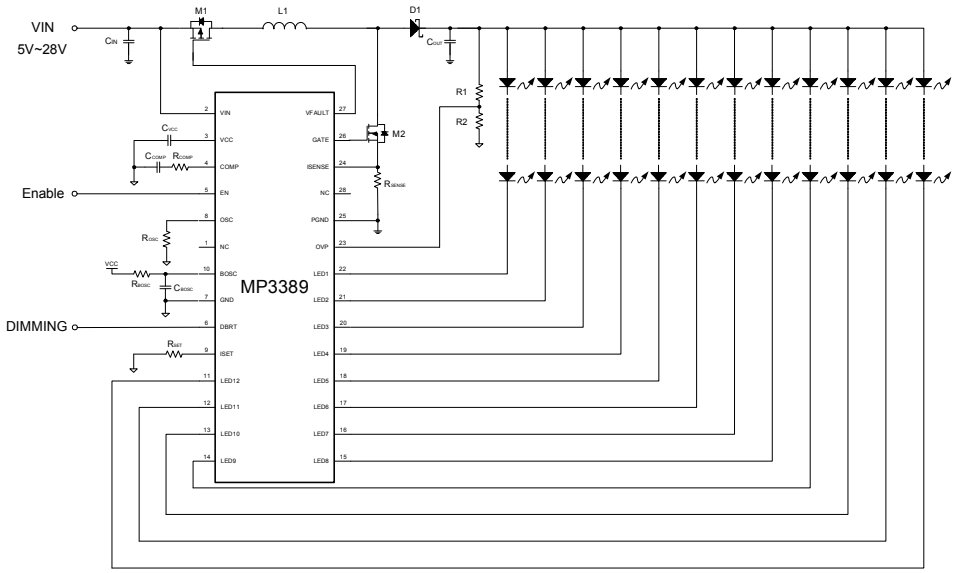
- High Efficiency and Small Size
- 5V to 28V Input Voltage Range
- Balanced Driver for 12 Strings of WLEDs
- Maximum 60mA for Each String
- 2.5% Current Matching Accuracy Between Strings
- Programmable Switching Frequency
- PWM or DC Input Burst PWM Dimming
- Open and Short LED protection
- Programmable Over-voltage Protection
- Under Voltage Lockout
- Thermal Shutdown
- 28-pin TSSOP and 28-pin SOIC Package

APPLICATIONS

- Desktop LCD Flat Panel Displays
- Flat Panel Video Displays
- LCD TVs and Monitors

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TYPICAL APPLICATION



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ORDERING INFORMATION

Part Number*	Package	Top Marking
MP3389EF	TSSOP28	3389EF
MP3389EY	SOIC28	3389EY

*For Tape & Reel, add suffix –Z (eg. MP3389EF–Z).

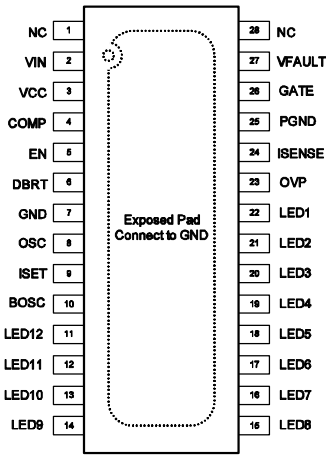
For RoHS compliant packaging, add suffix –LF (eg. MP3389EF–LF–Z)

**For Tape & Reel, add suffix –Z (eg. MP3389EY–Z).

For RoHS compliant packaging, add suffix –LF (eg. MP3389EY–LF–Z)

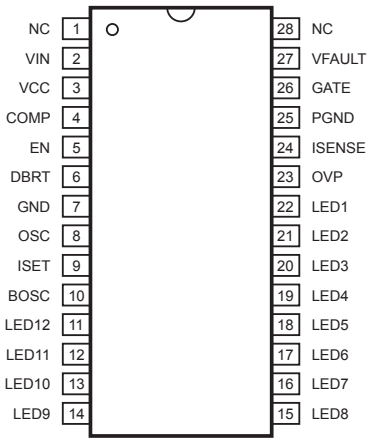
PACKAGE REFERENCE

TOP VIEW



TSSOP28

TOP VIEW



SOIC28

ABSOLUTE MAXIMUM RATINGS (1)

V_{IN} .....-0.3V to +30V
V_{VFAULT} ..... V_{IN} - 6V to V_{IN}
V_{GATE} .....-0.5V to 6.3V
V_{LED1} to V_{LED12} .....-1V to +50V
All Other Pins.....-0.3V to +6.3V
Continuous Power Dissipation (T_A = +25°C) (2)
TSSOP28…… ...3.9W
SOIC28…… ...2.1W
Junction Temperature.....150°C
Lead Temperature .....260°C
Storage Temperature..... -65°C to +150°C

Recommended Operating Conditions (3)

Supply Voltage V_{IN} .....5V to 28V
LED Current (Backlight) .....10mA to 60mA
Operating Junction Temp. (T_J). -40°C to +125°C

Thermal Resistance (4) θ_{JA} θ_{JC}

TSSOP28.....32 ..... 6 .... °C/W
SOIC28 .....60 ..... 30 ... °C/W

Notes:

1) Exceeding these ratings may damage the device.
2) The maximum allowable power dissipation is a function of the maximum junction temperature T_J (MAX), the junction-to-ambient thermal resistance θ_{JA}, and the ambient temperature T_A. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P_D (MAX) = (T_J (MAX)-T_A)/θ_{JA}. Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
3) The device is not guaranteed to function outside of its operation conditions.
4) Measured on JESD51-7, 4-layer PCB.

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ELECTRICAL CHARACTERISTICS

V_{IN} =12V, V_{EN} = 5V, T_A = +25°C, unless otherwise noted.

Parameters	Symbol	Condition	Min	Typ	Max	Units
Operating Input Voltage	V_{IN}		4.5		28	V
Supply Current (Quiescent)	I_Q	V_{IN}=12V, V_{EN}=5V, no load with switching		4		mA
Supply Current (Shutdown)	I_{ST}	V_{EN}=0V, V_{IN}=12V			2	μA
LDO Output Voltage	V_{CC}	V_{EN}=5V, 6V<V_{IN}<28V, 0<I_{VCC}<10mA	4.5	5	5.5	V
Input UVLO Threshold	V_{IN_UVLO}	Rising Edge	3.4	3.9	4.3	V
Input UVLO Hysteresis				200		mV
EN High Voltage	V_{EN_HIGH}	V_{EN} Rising	1.6			V
EN Low Voltage	V_{EN_LOW}	V_{EN} Falling			0.6	V
STEP-UP CONVERTER						
Gate Driver Impedance (Sourcing)		V_{CC}=5V,V_{GATE}=5V		4		Ω
Gate Driver Impedance (Sinking)		V_{CC}=5V,I_{GATE}=10mA		2		Ω
Switching Frequency	f_{SW}	R_{OSC}= 115kΩ	530	590	650	kHz
		R_{OSC}= 374kΩ	160	180	200	kHz
OSC Voltage	V_{OSC}		1.18	1.23	1.28	V
Minimum On Time	T_{ON_MIN}	PWM Mode, when no pulse skipping happens		100		ns
Maximum Duty Cycle	D_{MAX}		90			%
ISENSE Limit		Max Duty Cycle	175	220	265	mV
Leading Edge Blanking Time (6)	T_{LEB}			250		ns
PWM DIMMING						
DBRT Leakage Current	I_{DBRT_LK}		-5		+5	μA
BOSC Frequency	F_{BOSC}	C_{BOSC}=2.2nF	1.2	1.6	2	kHz
BOSC Output Current	I_{BOSC}		6.37	7.5	8.63	μA
LED CURRENT REGULATION						
ISET Voltage	V_{ISET}		1.20	1.22	1.245	V
LEDX Average Current	I_{LED}	R_{ISET}=40kΩ	29.6	30.5	31.4	mA
Current Matching (5)		I_{LED}=30.5mA			2.5	%
LEDX Regulation Voltage	V_{LEDX}	I_{LED}=30.5mA	410	550	690	mV
PROTECTION						
OVP Over Voltage Threshold	V_{OVP_OV}	Rising Edge	1.20	1.23	1.26	V

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ELECTRICAL CHARACTERISTICS *(continued)*

V_{IN} =12V, V_{EN} = 5V, T_A = +25°C, unless otherwise noted.

Parameters	Symbol	Condition	Min	Typ	Max	Units
OVP UVLO threshold	V_{OVP_UV}	Step-up Converter Fails	50	70	90	mV
LEDX Over Voltage Threshold	V_{LEDX_OV}	V_{IN}>5.5V	5.1	5.5	5.9	V
LEDX UVLO Threshold	V_{LEDX_UV}		140	180	220	mV
Thermal Shutdown Threshold	T_{ST}			130		℃
VFAULT Pull Down Current	I_{FAULT}		40	55	70	μA
VFAULT Blocking-Off Voltage (with Respect to V_{IN})	V_{FAULT}	V_{IN} =12V, V_{IN}-V_{FAULT}	5.3	5.8	6.3	V

Notes:

5) Matching is defined as the difference of the maximum to minimum current divided by 2 times average currents.
6) Guarantee by design.

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PIN FUNCTIONS

Pin #	Name	Description
1	NC	No Connect.
2	VIN	Supply Input. VIN supplies the power to the chip, as well as the step-up converter switch. Drive VIN with a 5V to 28V power source. Must be locally bypassed.
3	VCC	The Internal 5V Linear Regulator Output. VCC provides power supply for the internal MOSFET switch gate driver and the internal control circuitry. Bypass VCC to GND with a ceramic capacitor.
4	COMP	Step-up Converter Compensation Pin. This pin is used to compensate the regulation control loop. Connect a capacitor or a series RC network from COMP to GND.
5	EN	Enable Control Input. Do not let this pin floating.
6	DBRT	Brightness Control Input. To use external PWM dimming mode, apply a PWM signal on this pin for brightness control. To use DC input PWM dimming mode, apply a DC voltage range from 0.2V to 1.2V on this pin linearly to set the internal dimming duty cycle from 0% to 100%. The MP3389 has positive dimming polarity on DBRT.
7	GND	Analog Ground.
8	OSC	Switching Frequency Set. Connect a resistor between OSC and GND to set the step-up converter switching frequency. The voltage at this pin is regulated to 1.23V. The clock frequency is proportional to the current sourced from this pin.
9	ISET	LED Current Set. Tie a current setting resistor from this pin to ground to program the current in each LED string. The MP3389 regulates the voltage across the current setting resistor. The regulation voltage is 1.22V. The proportion of the current through the ISET resistor and the LED current is 1:1000.
10	BOSC	Dimming Repetition Set. This is the timing pin for the oscillator to set the dimming frequency. To use DC input PWM dimming mode, connect a capacitor from this pin to GND to set the internal dimming frequency. A saw-tooth waveform is generated on this pin. To use external PWM dimming mode, connect a 100kΩ resistor from this pin to GND, and apply the PWM signal on DBRT pin.
11	LED12	LED String 12 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 12 cathode to this pin. If this string is not used, connect Vin to this pin.
12	LED11	LED String 11 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 11 cathode to this pin. If this string is not used, connect Vin to this pin.
13	LED10	LED String 10 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 10 cathode to this pin. If this string is not used, connect Vin to this pin.
14	LED9	LED String 9 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 9 cathode to this pin. If this string is not used, connect Vin to this pin.

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PIN FUNCTIONS *(continued)*

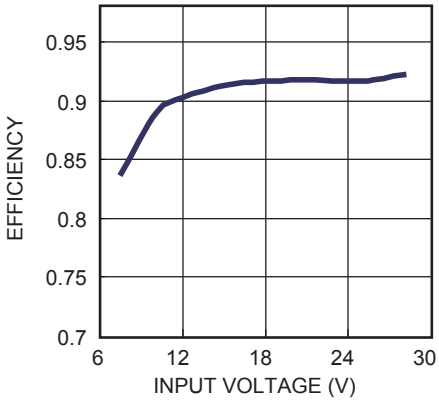
Pin #	Name	Description
15	LED8	LED String 8 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 8 cathode to this pin. If this string is not used, connect Vin to this pin.
16	LED7	LED String 7 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 7 cathode to this pin. If this string is not used, connect Vin to this pin.
17	LED6	LED String 6 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 6 cathode to this pin. If this string is not used, connect Vin to this pin.
18	LED5	LED String 5 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 5 cathode to this pin. If this string is not used, connect Vin to this pin.
19	LED4	LED String 4 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 4 cathode to this pin. If this string is not used, connect Vin to this pin.
20	LED3	LED String 3 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 3 cathode to this pin. If this string is not used, connect Vin to this pin.
21	LED2	LED String 2 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 2 cathode to this pin. If this string is not used, connect Vin to this pin.
22	LED1	LED String 1 Current Input. This pin is the open-drain output of an internal dimming control switch. Connect the LED String 1 cathode to this pin. If this string is not used, connect Vin to this pin.
23	OVP	Over-voltage Protection Input. Connect a resistor divider from output to this pin to program the OVP threshold. When this pin voltage reaches 1.23V, the MP3389 triggers OV Protection mode.
24	ISENSE	Current Sense Input. During normal operation, this pin senses the voltage across the external inductor current sensing resistor for peak current mode control and also to limit the inductor current during every switching cycle.
25	PGND	Step-up Converter Power Ground.
26	GATE	Step-up Converter Power Switch Gate Output. This pin drives the external power N-MOS device.
27	VFAULT	Fault Disconnection Switch Gate Output. When the system starts up normally, this pin smoothly turns on the external PMOS. When the MP3389 is disabled, the external PMOS is turned off to disconnect the input and output.
28	NC	No Connect.

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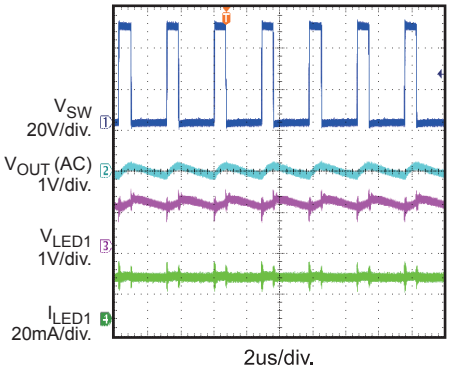
TYPICAL PERFORMANCE CHARACTERISTICS

V_{IN} =12V, 14 LEDs in series, 12 strings parallel, 20mA/string, unless otherwise noted.

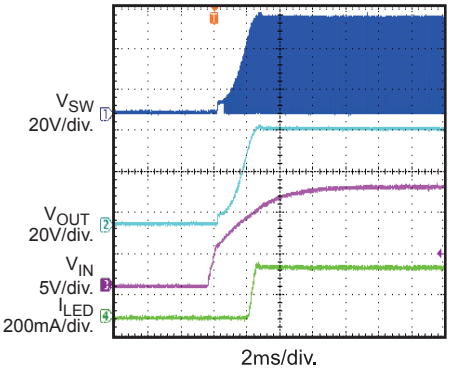
Efficiency vs. Input Voltage



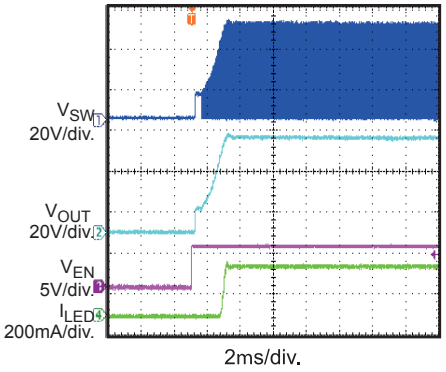
Steady State



Vin Startup

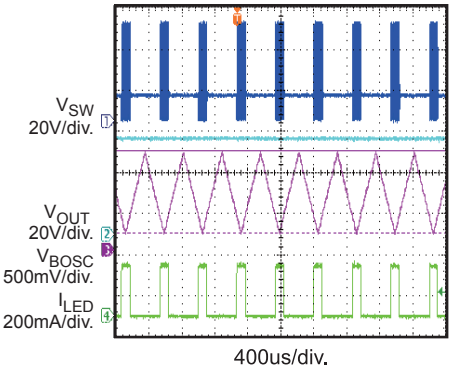


Ven Startup



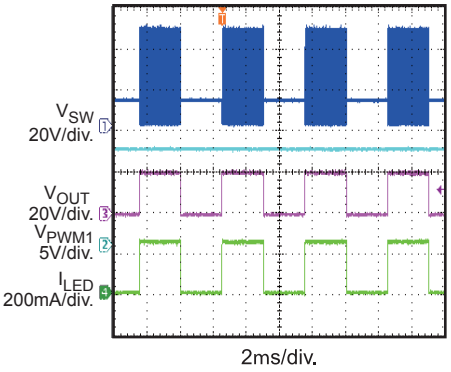
DC Burst Dimming

V_{PWM1} = 0.4V,C_{BOSC} = 1.8nF



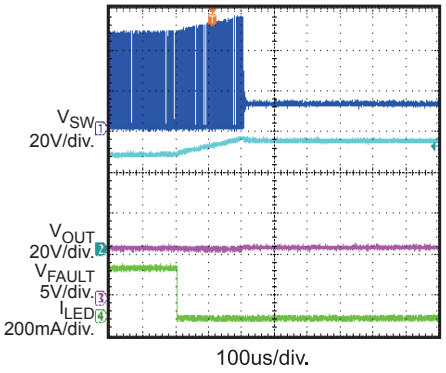
External PWM Dimming

f_{PWM} = 2kHz, D_{PWM} = 50%



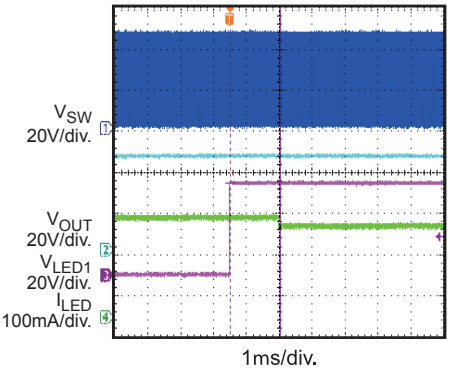
Open LED Protection

open all LED strings at working



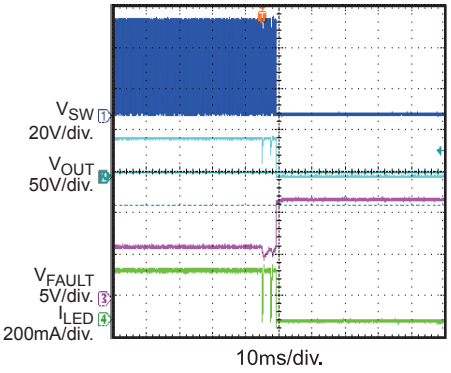
Short LED Protection

short V_{OUT} to LEDx at working



Short LED Protection

short V_{OUT} to GND at working



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FUNCTION DIAGRAM

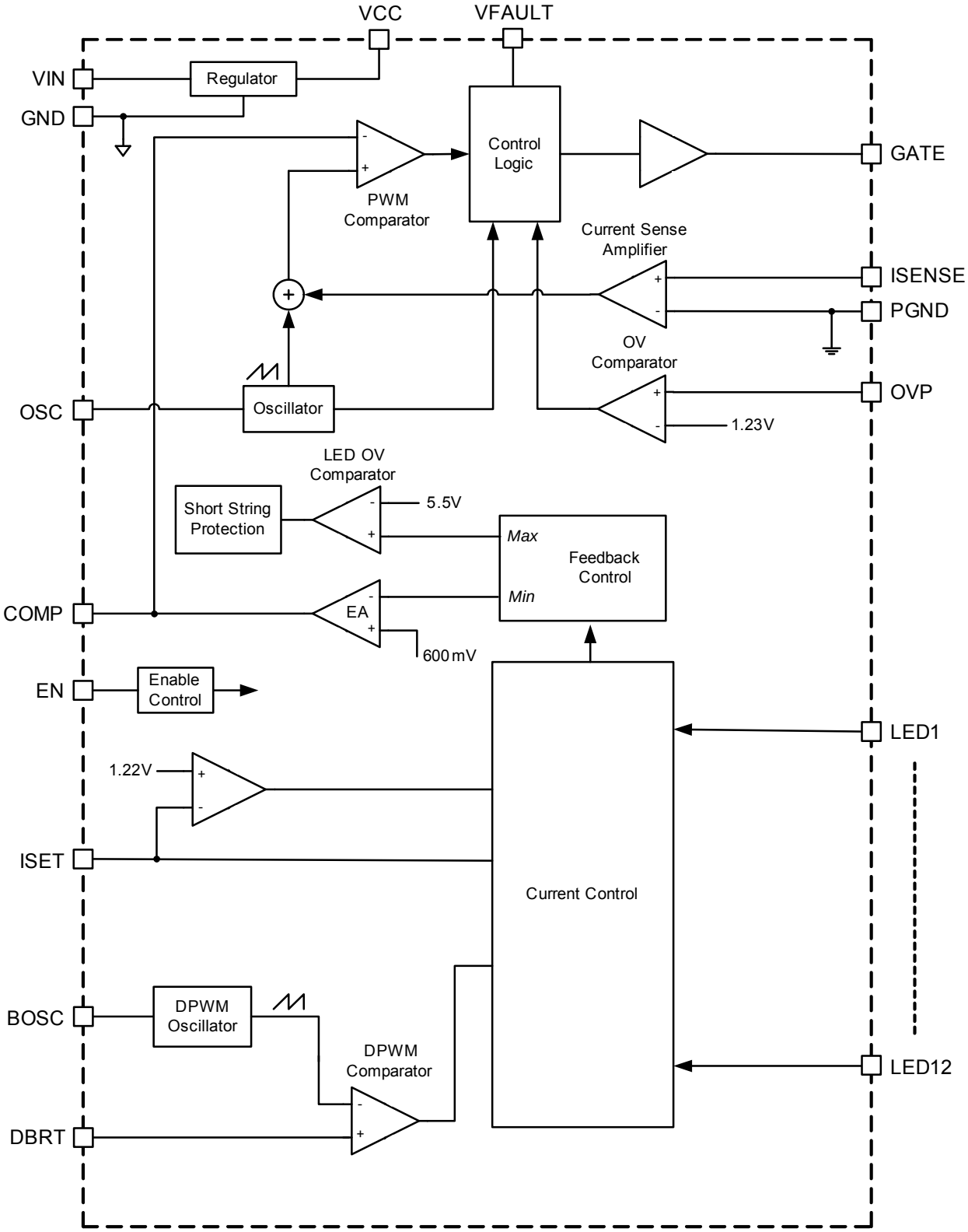


Figure 1—MP3389 Function Block Diagram

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OPERATION

The MP3389 employs a programmable constant frequency, peak current mode step-up converter and 12-channels regulated current sources to regulate the array of 12 strings white LEDs. The operation of the MP3389 can be understood by referring to the block diagram of Figure 1.

Internal 5V Regulator

The MP3389 includes an internal linear regulator (VCC). When VIN is greater than 5.5V, this regulator offers a 5V power supply for the internal MOSFET switch gate driver and the internal control circuitry. The VCC voltage drops to 0V when the chip shuts down. In the application of VIN smaller than 5.5V, tie VCC and VIN together. The MP3389 features Under Voltage Lockout. The chip is disabled until VCC exceeds the UVLO threshold. And the hysteresis of UVLO is approximately 200mV.

System Startup

When the MP3389 is enabled, the chip checks the topology connection first. The VFAULT pin drives the external Fault Disconnection PMOS to turn on slowly. Then the chip monitors the OVP pin to see if the Schottky diode is not connected or the boost output is short to GND. If the OVP voltage is lower than 70mV, the chip will be disabled and the external PMOS is turned off together. The MP3389 will also check other safety limits, including UVLO and OTP after the OVP test is passed. If they are all in function, it then starts boosting the step-up converter with an internal soft-start.

It is recommended on the start up sequence that the enable signal comes after input voltage and PWM dimming signal established.

Step-up Converter

The converter operation frequency is programmable (from 100kHz to 500kHz) with a external set resistor on OSC pin, which is helpful for optimizing the external components sizes and improving the efficiency.

At the beginning of each cycle, the external MOSFET is turned with the internal clock. To prevent sub-harmonic oscillations at duty cycles greater than 50 percent, a stabilizing ramp is added to the output of the current sense amplifier and the result is fed into the PWM comparator. When this result voltage reaches the output voltage of the error amplifier (V_{COMP}) the external MOSFET is turned off.

The voltage at the output of the internal error amplifier is an amplified signal of the difference between the 550mV reference voltage and the feedback voltage. The converter automatically chooses the lowest active LEDX pin voltage for providing enough bus voltage to power all the LED arrays.

If the feedback voltage drops below the 550mV reference, the output of the error amplifier increases. It results in more current flowing through the power FET, thus increasing the power delivered to the output. In this way it forms a close loop to make the output voltage in regulation.

At light-load or Vout near to Vin operation, the converter runs into the pulse-skipping mode, the FET is turned on for a minimum on-time of approximately 100ns, and then the converter discharges the power to the output in the remain period. The external MOSFET will keep off until the output voltage needs to be boosted again.

Dimming Control

The MP3389 provides two PWM dimming methods: external PWM signal or DC input PWM Dimming mode (see Figure 2). Both methods results in PWM chopping of the current in the LEDs for all 12 channels to provide LED control.

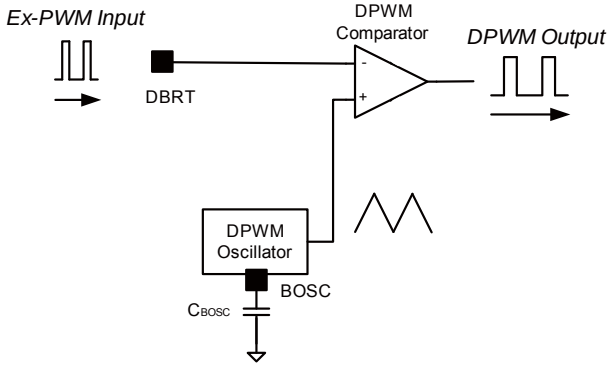


Figure 2—PWM Dimming Method

When bias the BOSC pin to a DC level, applying a PWM signal to the DBRT pin to achieve the PWM dimming. A DC analog signal can be

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directly applied to the DBRT pin to modulate the LED current. And the DC signal is then converted to a DPWM dimming signal at the setting oscillation frequency.

The brightness of the LED array is proportional to the duty cycle of the DPWM signal. The DPWM signal frequency is set by the cap at the BOSC pin.

Open String Protection

The open string protection is achieved through the over voltage protection. If one or more strings are open, the respective LEDX pins are pulled to ground and the IC keeps charging the output voltage until it reach OVP threshold. Then the part will mark off the open strings whose LEDX pin voltage is less than 180mV. Once the mark-off operation completes, the remaining LED strings will force the output voltage back into tight regulation. The string with the highest voltage drop is the ruling string during output regulation.

The MP3389 always tries to light at least one string and if all strings in use are open, the MP3389 shuts down the step-up converter. The part will maintain mark-off information until the part shuts down.

Short String Protection

The MP3389 monitors the LEDX pin voltage to judge if the short string occurs. If one or more strings are short, the respective LEDX pins will be pulled up to the boost output and tolerate high voltage stress. If the LEDX pin voltage is higher than 5.5V, the short string condition is detected on the respective string. When the short string fault (LEDX over-voltage fault) continues for greater than 512 switching clocks, the string is marked off and disabled. Once a string is marked off, its current regulation is forced to disconnect from the output voltage loop regulation. The marked-off LED strings will be shut off totally until the part restarts. If all strings in use are short, the MP3389 will shut down the step-up converter.

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APPLICATION INFORMATION

Selecting the Switching Frequency

The switching frequency of the step-up converter is programmable from 100kHz to 500kHz. A oscillator resistor on OSC pin sets the internal oscillator frequency for the step-up converter according to the equation:

$$f_{SW} = 67850 / R_{OSC} \text{（kΩ）}$$

For R_{OSC}=191kΩ, the switching frequency is set to 355 kHz.

Setting the LED Current

The LED string currents are identical and set through the current setting resistor on the ISET pin.

$$I_{LED} = 1000 \times 1.22V / R_{SET}$$

For R_{SET}=60.4kΩ, the LED current is set to 20mA. The ISET pin can not be open.

Selecting the Input Capacitor

The input capacitor reduces the surge current drawn from the input supply and the switching noise from the device. The input capacitor impedance at the switching frequency should be less than the input source impedance to prevent high frequency switching current from passing through the input. Ceramic capacitors with X5R or X7R dielectrics are highly recommended because of their low ESR and small temperature coefficients. For most applications, a 4.7μF ceramic capacitor paralleled a 220uF electrolytic capacitor is sufficient.

Selecting the Inductor and Current Sensing Resistor

The inductor is required to force the higher output voltage while being driven by the input voltage. A larger value inductor results in less ripple current, resulting in lower peak inductor current and reducing stress on the internal N-Channel MOSFET. However, the larger value inductor has a larger physical size, higher series resistance, and lower saturation current.

Choose an inductor that does not saturate under the worst-case load conditions. A good rule for determining the inductance is to allow the peak-to-peak ripple current to be approximately 30% to 40% of the maximum input current. Calculate the required inductance value by the equation:

$$L = \frac{V_{IN} \times (V_{OUT} - V_{IN})}{V_{OUT} \times f_{SW} \times \Delta I}$$

$$I_{IN(MAX)} = \frac{V_{OUT} \times I_{LOAD(MAX)}}{V_{IN} \times \eta}$$

$$\Delta I = (30\%\sim40\%) \times I_{IN(MAX)}$$

Where V_{IN} is the minimum input voltage, f_{SW} is the switching frequency, $I_{LOAD(MAX)}$ is the maximum load current, ΔI is the peak-to-peak inductor ripple current and ηis the efficiency.

The switch current is usually used for the peak current mode control. In order to avoid hitting the current limit, the voltage across the sensing resistor R_{SENSE} should be less than 80% of the worst case current limit voltage, V_{SENSE}.

$$R_{SENSE} = \frac{0.8 \times V_{SENSE}}{I_{L(PEAK)}}$$

Where $I_{L(PEAK)}$ is the peak value of the inductor current. V_{SENSE} is shown in Figure 3.

V_{SENSE} vs. Duty Cycle

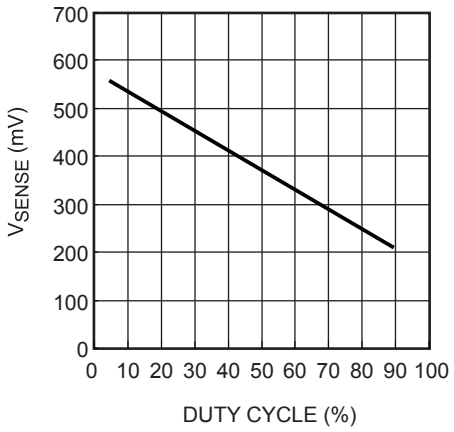


Figure 3—V_{SENSE} vs Duty Cycle

Selecting the Power MOSFET

The MP3389 is capable of driving a wide variety of N-Channel power MOSFETS. The critical parameters of selection of a MOSFET are:

1. Maximum drain to source voltage, $V_{DS(MAX)}$
2. Maximum current, $I_{D(MAX)}$
3. On-resistance, $R_{DS(ON)}$
4. Gate source charge Q_{GS} and gate drain charge Q_{GD}
5. Total gate charge, Q_G

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Ideally, the off-state voltage across the MOSFET is equal to the output voltage. Considering the voltage spike when it turns off, $V_{DS(MAX)}$ should be greater than 1.5 times of the output voltage.

The maximum current through the power MOSFET happens when the input voltage is minimum and the output power is maximum. The maximum RMS current through the MOSFET is given by

$$I_{RMS(MAX)} = I_{IN(MAX)} \times \sqrt{D_{MAX}}$$

Where:

$$D_{MAX} \approx \frac{V_{OUT} - V_{IN(MIN)}}{V_{OUT}}$$

The current rating of the MOSFET should be greater than 1.5 times $I_{RMS,}$

The on resistance of the MOSFET determines the conduction loss, which is given by:

$$P_{cond} = I_{RMS}{}^2 \times R_{DS(on)} \times k$$

Where k is the temperature coefficient of the MOSFET.

The switching loss is related to Q_{GD} and Q_{GS1} which determine the commutation time. Q_{GS1} is the charge between the threshold voltage and the plateau voltage when a driver charges the gate, which can be read in the chart of V_{GS} vs. Q_G of the MOSFET datasheet. Q_{GD} is the charge during the plateau voltage. These two parameters are needed to estimate the turn on and turn off loss.

$$P_{SW} = \frac{Q_{GS1} \times R_G}{V_{DR} - V_{TH}} \times V_{DS} \times I_{IN} \times f_{SW} + \frac{Q_{GD} \times R_G}{V_{DR} - V_{PLT}} \times V_{DS} \times I_{IN} \times f_{SW}$$

Where V_{TH} is the threshold voltage, V_{PLT} is the plateau voltage, R_G is the gate resistance, V_{DS} is the drain-source voltage. Please note that the switching loss is the most difficult part in the loss estimation. The formula above provides a simple physical expression. If more accurate estimation is required, the expressions will be much more complex.

For extended knowledge of the power loss estimation, readers should refer to the book "Power MOSFET Theory and Applications" written by Duncan A. Grant and John Gowar.

The total gate charge, Q_G, is used to calculate the gate drive loss. The expression is

$$P_{DR} = Q_G \times V_{DR} \times f_{SW}$$

Where V_{DR} is the drive voltage.

Selecting the Output Capacitor

The output capacitor keeps the output voltage ripple small and ensures feedback loop stability. The output capacitor impedance should be low at the switching frequency. Ceramic capacitors with X7R dielectrics are recommended for their low ESR characteristics. For most applications, a 4.7μF ceramic capacitor paralleled 10uF electrolytic capacitor will be sufficient.

Setting the Over Voltage Protection

The open string protection is achieved through the over voltage protection (OVP). In some cases, an LED string failure results in the feedback voltage always zero. The part then keeps boosting the output voltage higher and higher. If the output voltage reaches the programmed OVP threshold, the protection will be triggered.

To make sure the chip functions properly, the OVP setting resistor divider must be set with a proper value. The recommended OVP point is about 1.2 times higher than the output voltage for normal operation.

$$V_{OVP}=1.23V*(R_1+R_2)/R_2$$

Selecting Dimming Control Mode

The MP3389 provides 2 different dimming methods

1. Direct PWM Dimming

An external PWM dimming signal is employed to achieve PWM dimming control. Connect a 100kΩ resistor from BOSC pin to GND and apply the 100Hz to 2kHz PWM dimming signal to DBRT pin. The minimum recommended amplitude of the PWM signal is 1.2V. The low level should less than 0.4V. (See Figure 4).

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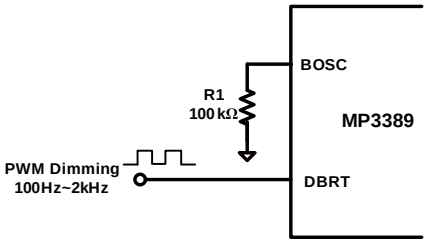


Figure 4—Direct PWM Dimming

Table 1 shows the PWM dimming duty Range with different PWM dimming frequency.

Tab 1 The Range of PWM Dimming Duty

fpwm(Hz)	Dmin	Dmax
100<f≤200	0.30%	100%
200<f≤500	0.75%	100%
500<f≤1k	1.50%	100%
1k<f≤2k	3.00%	100%
2k<f≤5k	7.50%	100%
5k<f≤10k	15.00%	100%
10k<f≤20k	30.00%	100%

2. DC Input PWM Dimming

To apply DC input PWM dimming, apply an analog signal (range from 0.2 V to 1.2V) to the DBRT pin to modulate the LED current directly. If the PWM is applied with a zero DC voltage, the PWM duty cycle will be 0%. If the DBRT pin is applied with a DC voltage>1.2V, the output will be 100% (See Figure 5). The capacitor on BOSC pin set the frequency of internal triangle waveform according to the equation:

$$F_{UNITE} = 3.5\mu F / C_{UNITE}$$

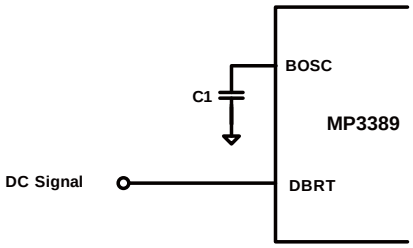


Figure 5—DC input PWM Dimming

Layout Considerations

Careful attention must be paid to the PCB board layout and components placement. Proper layout of the high frequency switching path is critical to prevent noise and electromagnetic interference problems. The loop of external MOSFET (M2), output diode (D1), and output capacitor (C5) is flowing with high frequency pulse current. it must be as short as possible (See Figure 6).

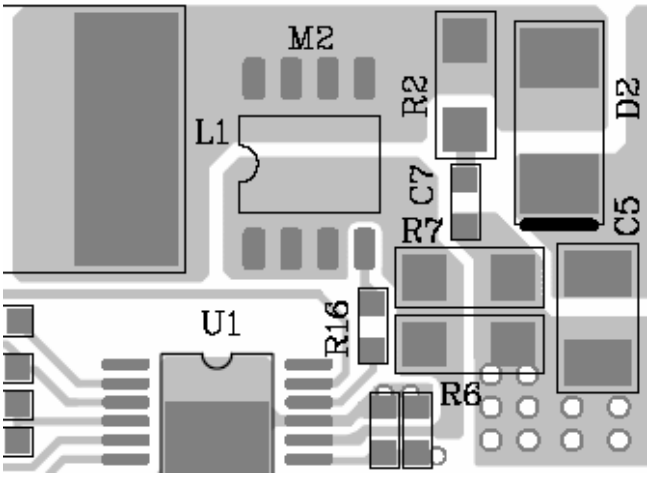


Figure 6—Layout Consideration

The IC exposed pad is internally connected to GND pin, and all logic signals are refer to the GND. The PGND should be externally connected to GND and is recommended to keep away from the logic signals.

External MOSFET for Short Protection

When output voltage is higher than rating of LED1-12 pins, external MOSFET is needed for each LED channel to avoid destroying LED1 -12 pins. (See Figure 8 for 4 channels application) The rating of MOSFET must be higher than 1.2* V_{OUT} t and more than 1.2* I_{LED}.

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TYPICAL APPLICATION CIRCUIT

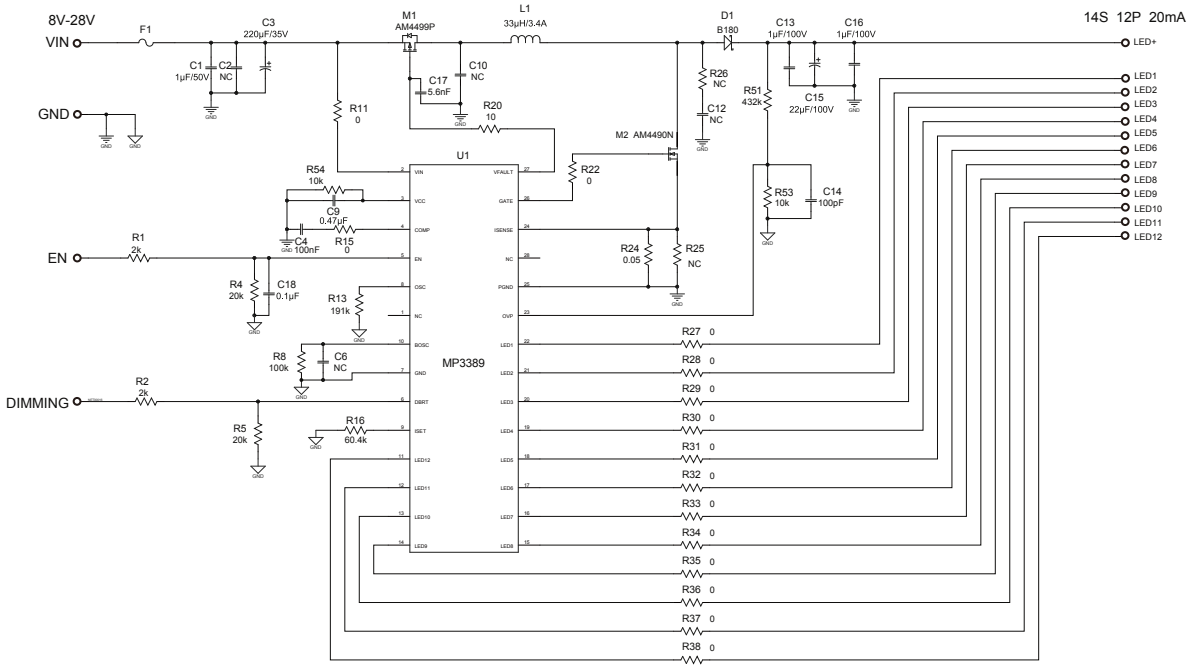


Figure 7—Drive 14 LEDs in Series, 12 Strings 20mA/string for Monitor Backlighting

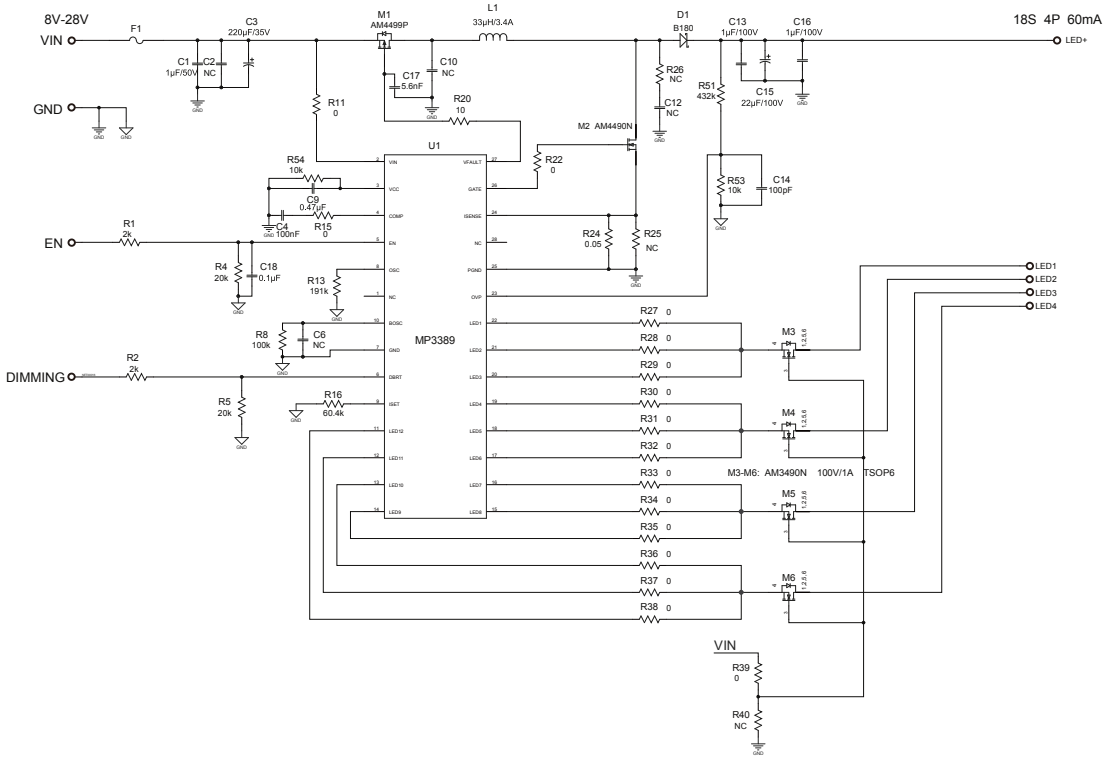


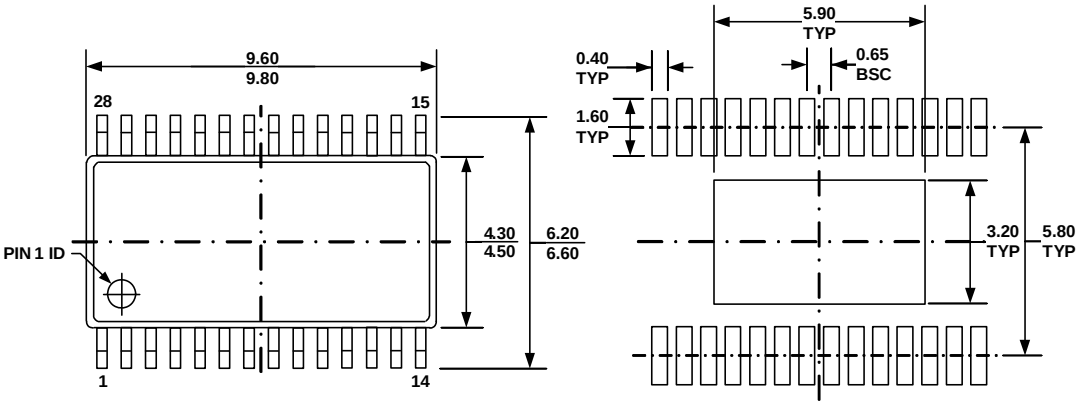
Figure 8 – Drive 18LEDs in Series, 4 strings 60mA/string

Note: For Vin>15V application, use R139 and R40(10kΩ) divider to make a 15V bias voltage for M3-M6.

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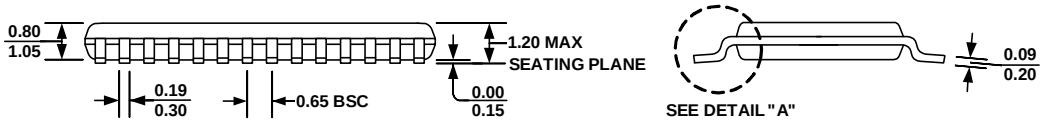
PACKAGE INFORMATION

TSSOP28



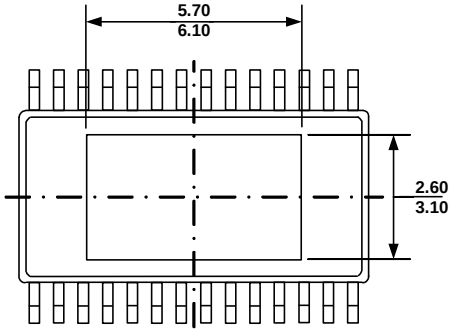
TOP VIEW

RECOMMENDED LAND PATTERN

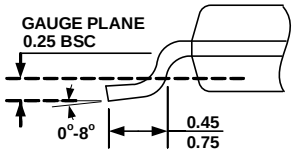


FRONT VIEW

SIDE VIEW



BOTTOM VIEW



DETAIL "A"

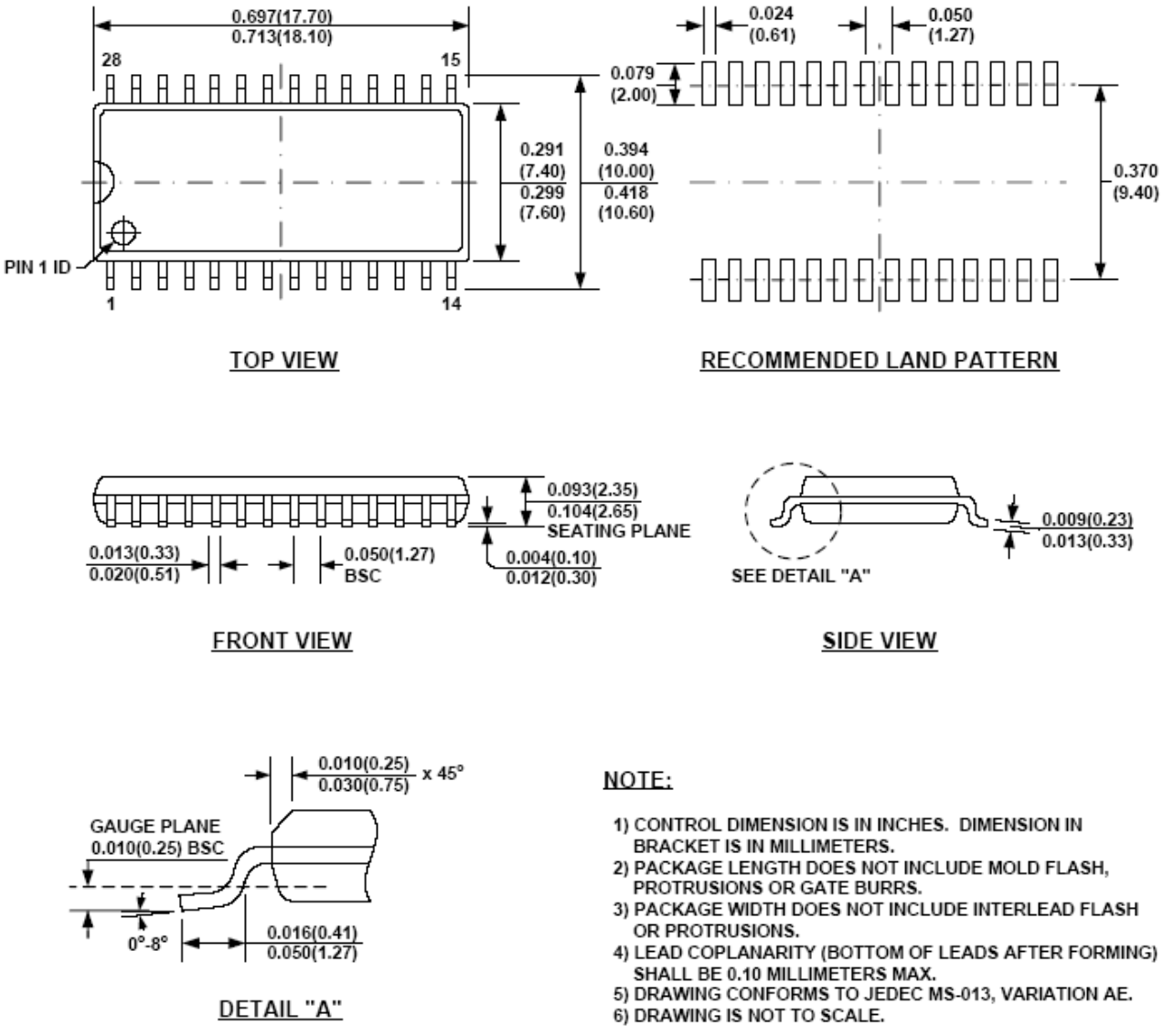
NOTE:

1) ALL DIMENSIONS ARE IN MILLIMETERS
2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSION OR GATE BURR
3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION
4) LEAD COPLANARITY(BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.10 MILLIMETERS MAX
5) DRAWING CONFORMS TO JEDEC MO-153, VARIATION AET.
6) DRAWING IS NOT TO SCALE

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PACKAGE INFORMATION

SOIC28



NOTE:

1) CONTROL DIMENSION IS IN INCHES. DIMENSION IN BRACKET IS IN MILLIMETERS.
2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
4) LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.10 MILLIMETERS MAX.
5) DRAWING CONFORMS TO JEDEC MS-013, VARIATION AE.
6) DRAWING IS NOT TO SCALE.

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